

# Design of a High Power MEMS Relay with Zero Voltage Switching and Isolated Power and Signal Transfer

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**Abstract**—This paper proposes a MEMS relay circuit by paralleling an auxiliary MOSFET with a MEMS switch. During the turn-on and turn-off transition, MOSFET is turned on at first to guarantee the switching voltage of MEMS switch is almost zero; for normal switch-on condition the MEMS switch with low on-resistance is turned on and the MOSFET is turned off. To meet the requirement of isolation between control side and power side, a transformer is designed in the control circuit and this single transformer achieves both power transfer and on-off control signal transfer. Thus the size of MEMS relay is reduced. Finally, a prototype was built and it validates the design of MEMS relay structure and single transformer circuit. Superior switching performance, fast response, low on-resistance and small size are achieved by the MEMS relay.

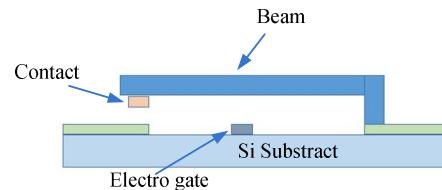
**Keywords**—MEMS relay; zero voltage switching; fast response;

## I. INTRODUCTION

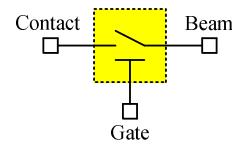
Electrical relay devices are widely used in any applications that require an isolated small signal to control high power circuit. The basic function of a relay is to switch on/off a high power circuit by an electrically isolated small power control signal. Electro-Mechanical Relay (EMR) driven by magnetic coils and Solid State Relay (SSR) based on silicon semiconductor device are the most commonly used types [1]-[3] in the industrial application. The former one is known for its low on-state resistance, high off-state voltage and large current capability while the latter for small size, mass producible, and easy on-chip integration. However, utilization of both types of relay has several obstacles. EMR suffers from the problems of mental-contact wear out, high voltage drive, and slow response time [4]. SSR has the relatively large on-resistance because it uses MOSFET as the conduction device and thus a large size of heat sink is needed to solve the thermal problem[5]. Furthermore, the opto-coupler which transfers the control signals in SSR limits the switching speed and slows the response.

MEMS switch is a device which has the advantages of SSR and EMR in low power applications. Until now, the great development of the metal device processing, high power system designs and the solid state micro switch fabrication

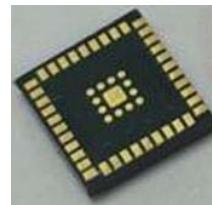
techniques significantly extends the power handling capability of MEMS switch [6][7]. Benefiting from this promising feature, MEMS switches can also be used in high power level. The existing MEMS switch with the minimum size of 5mm\*5mm QFG package is shown in Fig.1. It is capable of carrying more than 3A current at 200V switching voltage while only consuming pico amperes of current. MEMS switch has the following obvious advantages in: a) low cost with surface micro-machining techniques, b) longer lifetime: more than 3 billion switching cycles; c) near zero power consumption and ultra-low insertion loss; d) easy to achieve multi-chips parallelism due to the positive temperature coefficient (on-state resistance of the MEMS switch with the higher current will



(a) Cross section of the physical structure.



(b) Single channel.



(c) Single chip.

Fig.1 MEMS switch structure, layout and prototype.

increase thus the current through paralleled MEMS switches will be balanced).

However, the main disadvantage is MEMS switch cannot withstand high voltage and current overlap during switching transition [8-15]. Charging and discharging can lead to arcing problems and this phenomenon is even more severe with high voltage and current. It may incur short-time temperature rise to melt or evaporate the contact, and even if the instant over heating does not occur, this energy will damage the MEMS device eventually [8] [9]. So MEMS switch cannot be directly used as the power relay. Traditional users of both EMR and SSR such as medical test and measurement equipment, instrumentation system, radar system and satellite communication system are not directly applicable.

In order to explore the promising feature of MEMS switch in the aforementioned high power application, this paper proposes a new MEMS relay circuit structure which parallels a MOSFET with the MEMS switch to achieve zero voltage switching of the MEMS switch. The switching energy is reduced which is critical to improve its reliability under high voltage and large current operations. This paper is organized as follows: the proposed circuit structure and the detailed operation principle are demonstrated in Section II. A single transformer isolated circuit is proposed to achieve the control signal transfer and driver voltage supply for MOSFET, MEMS switch in section III. Then, a MEMS relay prototype is built to validate the high voltage-high current switching capability and the superior performance of MEMS switch. The waveforms of MEMS on-off process and over current protection mode are illustrated in section IV. Section V draws the conclusion.

## II. THE STRUCTURE OF MEMS RELAY

Despite the promising performance, MEMS switch is still easy to be damaged during switching transition if the voltage across the device is not zero or close to zero. In contrast to the MEMS switch, silicon based MOSFET can handle large voltage and current overlap during the switching transition. However, the main disadvantage of MOSFET is relatively larger on-state resistance. The MEMS relay is proposed by paralleling a MOSFET with the MEMS switch, the main structure is shown in Fig. 2. An auxiliary high voltage high current and low-cost MOSFET is introduced to be in parallel with the MEMS switch to protect the MEMS switch. It is turned on before the turn on and turn off transition of MEMS switch thus provides almost zero voltage commutation. In consequence the MEMS relay circuit combines the advantages of both MOSFET and MEMS switch and it can replace EMR and SSR in many applications to meet the same circuit requirement and achieve much better performance.

An isolated circuit is designed to generate the gate driving voltage for MEMS switch and MOSFET from a 5V power supply which also serves as the on-off control voltage. The outputs of the isolated driver include a 75V high voltage gate driver for MEMS switch and a 12V low voltage driver for MOSFET. It also transfers the control signal  $v_{con}$  from the

power side to the control side and generates the gate signal  $G_{MOS}$  and  $G_{MEMS}$ .

The key waveforms of turn on and turn off operation of MEMS switch based relay with parallel auxiliary MOSFET are illustrated in Fig.3.  $v_{con}$ ,  $G_{MOS}$  and  $G_{MEMS}$  are the on-off control and gate signals of the relay, MOSFET and MEMS switch, respectively. According to the different switching state, there are 5 operation modes during each switching commutation. Before MEMS relay is turned on, both MEMS switch and MOSFET are in off state, and the load current  $i_L$  is zero.

### A. Switching On Time Sequence

Mode 1 ( $t_0-t_1$ ): at  $t_0$ , the turn-on control signal  $v_{con}$  is step up. After a little time delay,  $G_{MOS}$  steps up immediately at  $t_1$  to turn on MOSFET first.

Mode 2 ( $t_1-t_2$ ): when MOSFET is turned on at  $t_1$ , the voltage across MEMS switch  $v_{PN}$  begins to decrease and the load current flows through MOSFET is increasing. At  $t_2$ ,  $v_{PN}$  decreases to zero and  $i_{MOS}$  increases to the load current  $i_{load}$ .

Mode 3 ( $t_2-t_3$ ): After a very small time delay until  $t_3$ ,  $i_{MOS}$  is

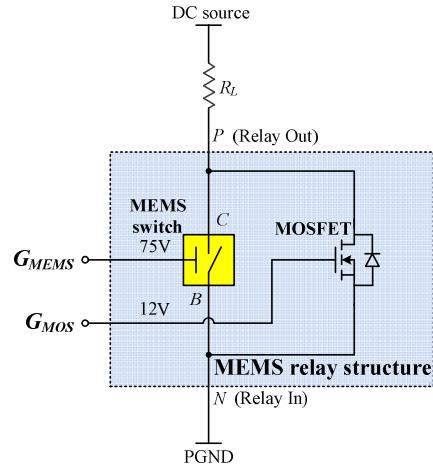


Fig.2 MEMS switch based relay with an auxiliary MOSFET.

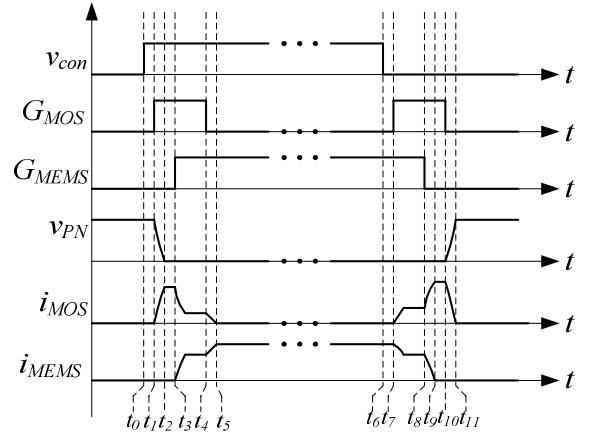


Fig.3 Operation principle of MEMS relay.

TABLE I Specification Comparison of Mechanical Relay, Solid State Relay and MEMS Relay of 200VDC rated voltage and 10A rated current.

Comparison items	EM Relay	Solid State Relay	MEMS Relay
On-state resistance	<100mΩ	<230mΩ	<50mΩ
ON/OFF switching time	>20ms	>1ms	<10us
Leakage current	75pA@200V	0.4mA@200V	75pA@200V
Switching operations	<30 million	<100 million	>3 billion

stabilized at the load current  $i_{MOS}=i_{load}$ .

Mode 4 ( $t_3-t_4$ ): at  $t_3$ ,  $v_{PN}$  is almost zero  $v_{PN}=0$ . MEMS switch is turned on.  $i_{MOS}$  begins to decreases and  $i_{MEMS}$  starts to increase. In this switching on process, MEMS switch achieves zero voltage turn on. Until  $t_4$ , MEMS switch is fully on state and the load current  $i_L$  is distributed in MEMS switch and MOSFET according to the on-state resistance.

Mode 5 ( $t_4-t_5$ ): at  $t_4$ , MOSFET is turned off.  $i_{MOS}$  is decreasing to zero. All the load current flows through MEMS switch until  $t_5$ .

### B. Switching OffTime Sequence

Mode 6 ( $t_6-t_7$ ): at  $t_6$ , the control signal  $v_{con}$  is coming to turn off the relay.  $G_{MOS}$  steps up immediately at  $t_7$  to turn on MOSFET first.

Mode 7 ( $t_7-t_8$ ): when MOSFET is turned on at  $t_7$ ,  $i_{MOS}$  begins to increases and  $i_{MEMS}$  starts to decreases. At  $t_8$ , Both MEMS switch and MOSFET are conducting load current  $i_L$ .

Mode 8 ( $t_8-t_9$ ): at  $t_8$ , MEMS switch is turned off.  $v_{PN}$  is almost zero  $v_{PN}=0$  because MOSFET is still conducting.  $i_{MEMS}$  decreases to zero at  $t_9$ . The load current is flowing through MOSFET. During this switching off process, MEMS switch achieves zero voltage turn off.

Mode 9 ( $t_9-t_{10}$ ): after a very small time delay until  $t_{10}$ ,  $i_{MOS}$  is stabilized at the load current  $i_{MOS}=i_{load}$ .

Mode 10 ( $t_{10}-t_{11}$ ): at  $t_{10}$ , MOSFET is turned off.  $i_{MOS}$  begins to decrease to zero and  $v_{PN}$  is increasing to the DC source voltage. At  $t_{11}$ , the relay is turned off. Selection of MOSFET

As an auxiliary component, the selection of MOSFET is a critical issue to guarantee that MEMS will be able to achieve the requirement of ZVS and also reduce the cost. Take consideration of the MEMS turn on process, a small enough  $R_{d\_MOS}$  should be selected to ensure the switching voltage of MEMS switch  $V_{MOS\_ON}$  is less than 0.5V to 1.0V at full load current after the MOSFET is turned on. The maximum on resistance follows the equation of(1):

$$R_{d\_MOS} = \frac{V_{MOS\_ON}}{I_{load\_max}} \quad (1)$$

Where  $V_{MOS\_ON}$  is MOSFET voltage drop under the maximum load current  $I_{Load\_max}$ .

However, it is not necessary to minimize the MOSFET on resistance. Firstly, the contact resistance of MEMS is already very small, selecting a small on-resistance value will not decrease  $i_{MEMS}$  in Mode 7. Also, during  $t_3-t_4$  and  $t_6-t_7$  interval, the load current is distributed in MEMS switch and MOSFET according to the on-state resistance. We want the most of current flows across MEMS switch to keep a constant operating condition, thus  $R_{d\_MOS}$  is supposed to be larger than  $R_{d\_MEMS}$ . This actually allows the cost to be lower.

Furthermore, the auxiliary MOSFET switch almost has no power loss and works in relatively low frequency, thus a small package and low-cost MOSFET can be used. With the established MEMS relay, comparisons are made as shown in Table 1 and it can achieve much better performance than conventional SSR and ESR devices.

### III. CONTROL CIRCUIT DESIGN WITH SINGLE TRANSFORMER

The compulsory isolation between control side and power side is required for MEMS relay [16-21]. Furthermore, the control circuit has the following requirements: a) transfer 5V on-off control command voltage to the MEMS switch and MOSFET gate signals in the secondary side. b) transfer 5V control voltage to 5V, 12V and 75V voltage in the secondary to provide power supply for control logic chip, MOSFET and MEMS switch driving circuit.

In order to minimize the relay size, an isolated circuit is proposed to acquire both secondary side power supply and on-off command information with single transformer. The diagram of control circuit is shown in Fig.4. As can be observed, when the turn-on control signal  $v_{con}=5V$  is applied, pulse generator (such as LMC555) generates the 500kHz pulse waveform and then this signal is power amplified by ADP3624 as the primary side voltage of transformer.  $C_1$  is used to filter the DC component. The voltage doubling rectifying circuit including  $D_1$ ,  $D_2$ ,  $C_2$ ,  $C_3$  is implemented to step up the transformer secondary voltage to the intermediate dc-link voltage  $V_{cc}=6V$  which serves as the input of voltage regulation circuit. A high step-up DC-DC chip is used to generate 75V voltage for MEMS switch. A voltage doubler chip is used to generate 12V for MOSFET drive. And a linear voltage regulator is used as 5V power supply for the control circuit.

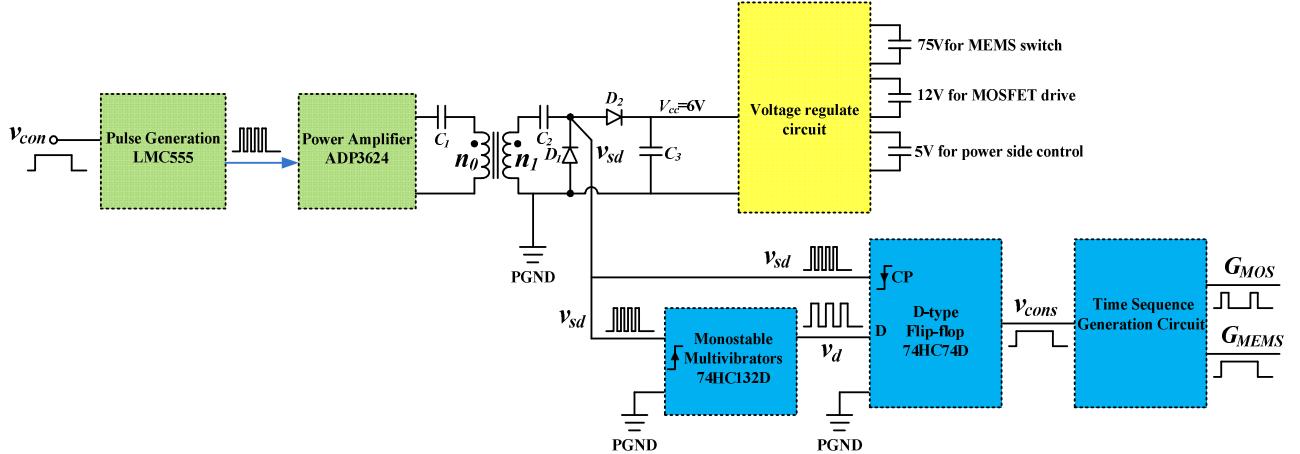


Fig.4 Control circuit diagram of MEMS relay.

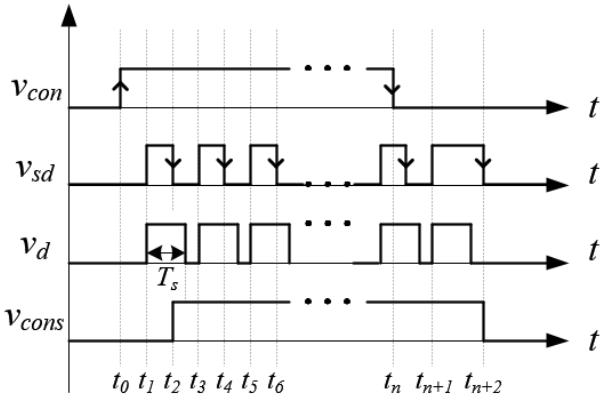


Fig.5 Operation principle of the control signal ( $v_{con}$ ) detection circuit.

The secondary side also provides the MEMS relay on-off signal  $v_{sd}$  to the control circuit. The MEMS switch on/off signal  $v_{cons}$  in the secondary side is generated by a D-type flip-flop 74HC74D.  $v_{sd}$  is the trigger source(falling edge trigger) of the flip-flop, and the input ( $v_d$ ) of the flip-flop is from the output of a monostable multivibrator 74HC132. Then, the time sequence circuit generates the gate control signal for MOSFET and MEMS switch according to operation principles of MEMS relay shown in Fig.3.

The detailed operation principle of control signal detection circuit is shown in Fig.5. Signals  $v_{sd}$  and  $v_d$  determine the on/off signal  $v_{cons}$  of MEMS switch by a D-type flip-flop, where  $v_d$  is the output of monostable multivibrator. The on-time of  $v_d$  is fixed ( $T_s \approx 1/2f_{s(min)} = 1.25\mu s$ ). Signals  $v_{sd}$  and  $v_d$  shares the same rising edge.

1) When control signal  $v_{con}$  is high ( $v_{con}=5V$ ), the frequency of  $v_{sd}$  is increased and its on-time is shorter than that of  $v_d$ . Therefore, when  $v_{sd}$  changes from high to low,  $v_d$  is still at high level. The falling edge of  $v_d$  triggers the D flip-flop so that the output of D flip-flop is always high ( $v_{cons}=5V$ ) when the control signal  $v_{con}$  is high.

2) When control signal  $v_{con}$  is low ( $v_{con}=0V$ ), the oscillating frequency of transformer decreases and the on-time of  $v_{sd}$  is longer than that of  $v_d$ . Therefore, when  $v_{sd}$  changes from high to low,  $v_d$  is low, and the output of D flip-flop goes to zero ( $v_{cons}=0$ ).

By using the proposed control circuit, the signal and power isolation are achieved with only one transformer and the size of MEMS relay can be reduced.

#### IV. EXPERIMENT VERIFICATION

##### A. Test on the single MEMS prototype

The prototype of MEMS relay shown in Fig.6 was built to verify the improved switching performance from our circuit design and theoretical analysis. The paralleled MOSFET is mounted at the back side of MEMS. The studied case for a single MEMS relay is 200V/3A power capacity as maximum. MOSFET 18N55M5 with 550V maximum drain to source voltage, 16A drain current and 192mΩ on resistance is selected for experiment. The test condition is:  $V_{dc}=50\sim 200V$ ,  $R_{Load}=25\sim 100\Omega$  to obtain load current with different voltage conditions. The on resistance of MEMS  $R_{d,MOS}$  is 70mΩ and the on resistance of selected auxiliary MOSFET is 192mΩ.

Fig.7 shows the waveforms of the MEMS relay during turn-on transition. The first channel is clock signal and its end time indicates the actual turn-on time of MEMS switch driver. The second channel shows the gate voltage of the MOSFET. The green curve is the total load current through the relay and the purple one is the voltage across the input and output terminals. It can be observed that the response time is only 30μs and the voltage has already reached almost zero before MEMS switch turning on. Fig. 8 shows the waveforms while the relay is turn off transition and the response time is even shorter (20μs). The 200V/2A on-off operation can be performed safely according to the figures. A snubber circuit is added to the prototype thus the current and voltage will change slowly after the MOSFET is turned off.

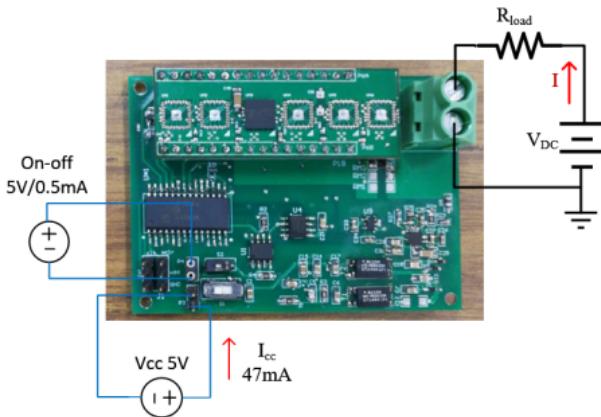


Fig.6 Prototype of MEMS relay

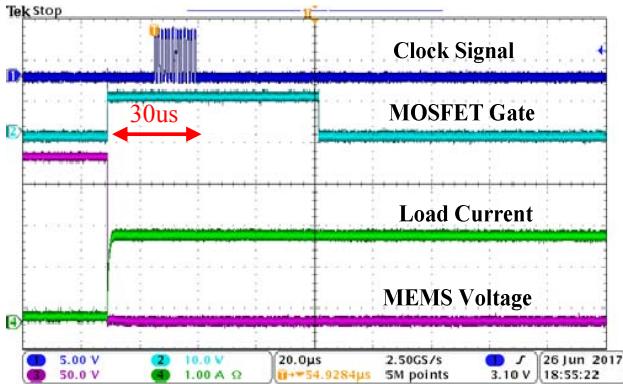


Fig.7 MEMS relay during turn-on transition.

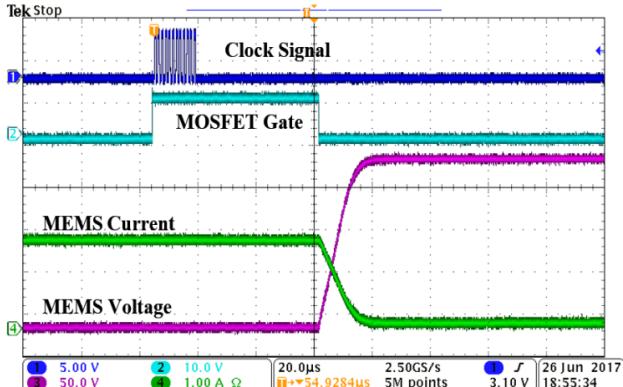


Fig.8 MEMS relay during turn-off transition.

Benefiting from the smart and fast response feature of the proposed MEMS relay, it can operate in switching mode at more than 1k Hz as shown in Fig.9. This feature also contributes to the fast over current protection, the MEMS switch can be shut down in a very short time when over current is detected. Fig. 10 and Fig. 11 demonstrate the waveforms in the protection procedure. The MOSFET is turned on as soon as load current hits the threshold as shown in Fig. 10 and then MEMS switch is turned off properly as the normal turn-off mode does.

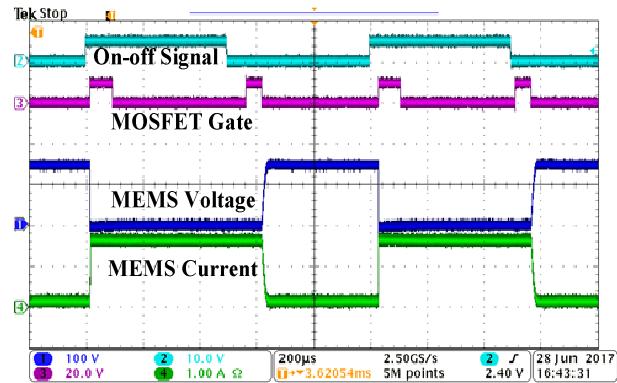


Fig.9 Signals of MEMS relay in on-off mode.

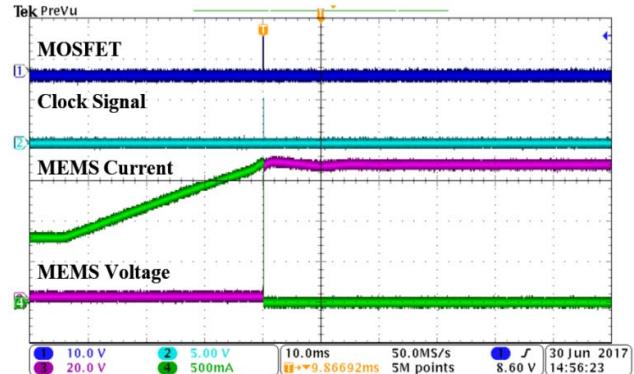


Fig.10 Over current protection in long time scale.

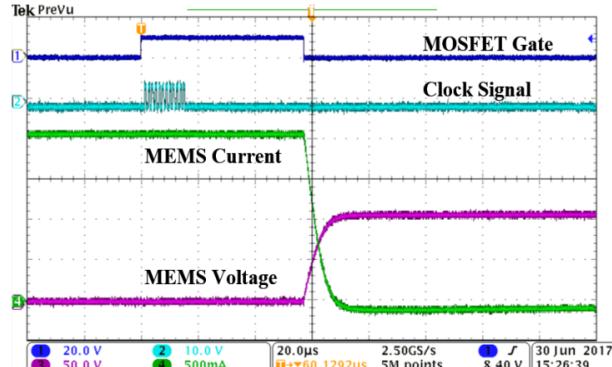


Fig.11 Over current protection in short time scale.

The voltage on the MEMS relay during the turn on and turn off transition are shown in Fig. 12 and Fig. 13. In period 1 of Fig. 12, when the MOSFET is on and MEMS is not on, the voltage drops to hundreds of millivolts; then during period 2, both MOSFET and MEMS are on and the voltage is a little lower; then in period 3 when the MOSFET is off MEMS voltage increases and reaches the input voltage. The oscillation occurs after MOSFET is turned on because large  $dv/dt$  is imposed and parasitic parameters draw this effect.

The voltage waveform is similar in turn off transition, it decreases in period 1 after the MOSFET is on and becomes

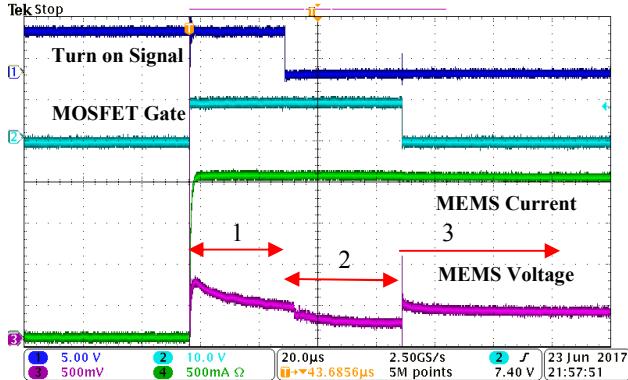


Fig.12 Detail MEMS during turn-on transition.

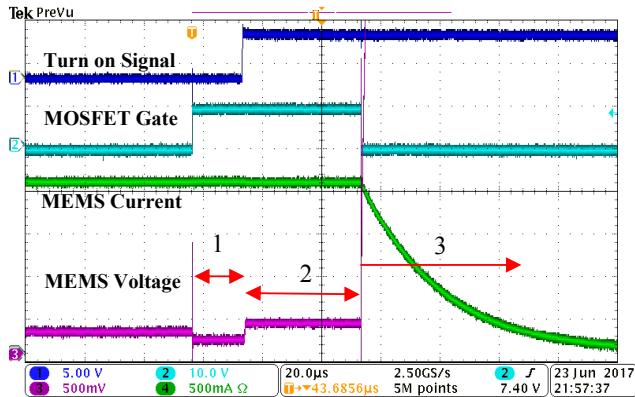


Fig.13 Detailed MEMS voltage during turn-off transition.

higher while the MEMS switch is off in period 2, then it rises after the MOSFET is turned off and the relay is off to shut down the power circuit. The oscillation is not as large as that in period 1, 2 and 3 in Fig. 12 because the  $dv/dt$  is much smaller thus the impact from parasitic parameters can be ignored.

#### B. Test on three-MEMS prototype

Another prototype with three MEMS switches and three MOSFETs in parallel is also built to test the performance with higher load current. The prototype is shown in Fig.14 and it is different from that in Fig. 6 by adding two pairs of MEMS switch and MOSFET on the top and bottom side of PCB. All the control signals are in parallel thus the three MOSFETs are turned on and off at the same time. Same condition happens to the MEMS switches.

It can be observed from Fig. 15 and Fig. 16 that the paralleled MEMS relays can be turned on and off safely with 4A load current at 100V input voltage. The MEMS switches are also turned on 30us after the  $v_{cons}$  high signal arrives and can stay stable after the MOSFETs are turned off. During the transition the voltage drop is less than 1V therefore the switching energy of MEMS is very small.

On the parallel MEMS relay prototype, full load operation up to 10A current and its over current protection will be further investigated. Also the current sharing and thermal distribution characteristics will be studied.



Fig.14 Three MEMS in parallel prototype.

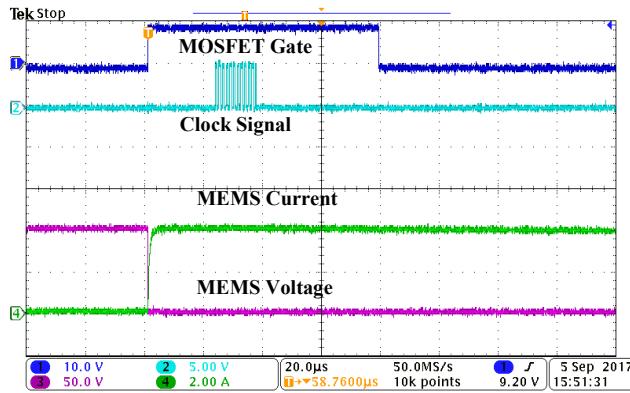


Fig.15 Waveforms during 3 MEMS turn-on transition.

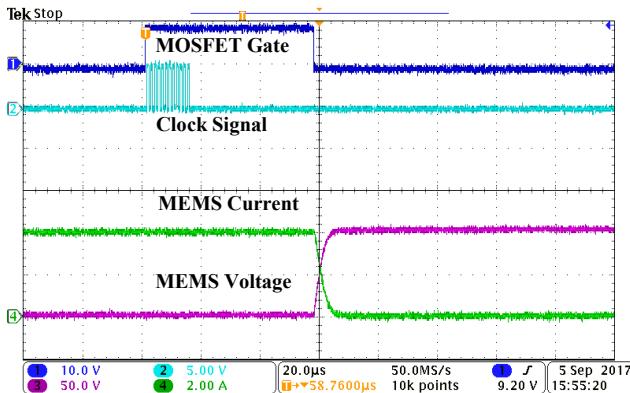


Fig.16 Waveforms during 3 MEMS relay turn-off transition.

#### V. CONCLUSION

The paper propose a novel MEMS based relay to achieve much better performance than conventional SSR and ESR. A paralleled auxiliary MOSFET is introduced to provide the zero voltage switching condition for MEMS switch. The ZVS condition significantly decrease the risk that may incur when the large voltage and current are overlapped during switching process. Furthermore, the control circuit uses only one transformer to minimize the relay size. Experimental results are performed and validate MEMS based relay can operate

under the serious switching condition with good performance. The voltage drop is less than 1V which effectively reduces the failure rate.

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